An Introduction to Information Theory

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Outline of the Talk

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- Fundamentals of Information Theory
- Research Directions: Memory Channels
- Research Directions: Continuous-Time Information Theory

Fundamentals of Information Theory

The Birth of Information Theory

Continuous-Time Information Theory 0000000

The Birth of Information Theory



The Birth of Information Theory



C. E. Shannon. A mathematical theory of communication. *Bell Syst. Tech. J.*, 27:379-423,623-656, 1948.

Classical Information Theory

Classical Information Theory



FIGURE 1.2. Information theory as the extreme points of communication theory.

Information Theory Nowadays



FIGURE 1.1. Relationship of information theory to other fields.

Entropy: Definition

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The entropy H(X) of a discrete random variable X is defined by

$$H(X) = -\sum_{x \in \mathcal{X}} p(x) \log p(x).$$

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Let

$$X = egin{cases} 1 & ext{ with probability } p, \ 0 & ext{ with probability } 1-p. \end{cases}$$

Then,

$$H(X) = -p \log p - (1-p) \log(1-p) \triangleq H(p).$$

Entropy: Measure of Uncertainty

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FIGURE 2.1. *H*(*p*) vs. *p*.

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Joint Entropy and Conditional Entropy

The joint entropy H(X, Y) of a pair of discrete random variables (X, Y) with a joint distribution p(x, y) is defined as

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$$H(Y|X) = \sum_{x \in \mathcal{X}} p(x)H(Y|X=x)$$

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All Entropies Together Chain Rule

H(X,Y) = H(X) + H(Y|X).

Chain Rule

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= $H(X) + H(Y|X).$

Original Definition

The mutual information I(X; Y) between two discrete random variables X, Y with joint distribution p(x, y) is defined as

$$I(X;Y) = \sum_{x,y} p(x,y) \log \frac{p(x,y)}{p(x)p(y)}.$$

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Alternative Definitions

$$I(X;Y) = H(X) + H(Y) - H(X,Y)$$

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Alternative Definitions

$$I(X; Y) = H(X) + H(Y) - H(X, Y) = H(X) - H(X|Y) = H(Y) - H(Y|X)$$

Mutual Information and Entropy

Mutual Information and Entropy



FIGURE 2.2. Relationship between entropy and mutual information.

Continuous-Time Information Theory 0000000

Asymptotic Equipartition Property Theorem

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AEP Theorem
f
$$X_1, X_2, ...,$$
 are i.i.d $\sim p(x)$, then
 $-\frac{1}{n} \log p(X_1, X_2, ..., X_n) \rightarrow H(X)$ in probability.
Asymptotic Equipartition Property Theorem

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Proof.

$$-\frac{1}{n}\log p(X_1, X_2, \dots, X_n) = -\frac{1}{n}\sum_{i=1}^n \log p(X_i)$$

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Proof.

$$-rac{1}{n}\log p(X_1,X_2,\ldots,X_n) = -rac{1}{n}\sum_{i=1}^n\log p(X_i)
ightarrow -\mathbb{E}[\log p(X)]$$

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 are i.i.d $\sim p(x)$, then
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Proof.

$$-\frac{1}{n}\log p(X_1,X_2,\ldots,X_n)=-\frac{1}{n}\sum_{i=1}^n\log p(X_i)\to -\mathbb{E}[\log p(X)]=H(X).$$

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where H(X) here denotes the entropy rate of the process $\{X_n\}$, namely,

$$H(X) = \lim_{n \to \infty} H(X_1, X_2, \ldots, X_n)/n.$$

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Proof.

There are many. The simplest is the **sandwich** argument by Algoet and Cover [1988].

Continuous-Time Information Theory

Typical Set: Definition and Properties

Definition The typical set $A_{\epsilon}^{(n)}$ with respect to p(x) is the set of sequence $(x_1, x_2, \ldots, x_n) \in \mathcal{X}^n$ with the property

$$H(X) - \epsilon \leq -\frac{1}{n} \log p(x_1, x_2, \dots, x_n) \leq H(X) + \epsilon.$$

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Properties

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Properties

•
$$(1-\varepsilon)2^{n(H(X)-\epsilon)} \le |A_{\epsilon}^{(n)}| \le 2^{n(H(X)+\epsilon)}$$
 for *n* sufficiently large.

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Properties

(1 − ε)2^{n(H(X)−ε)} ≤ |A_ε⁽ⁿ⁾| ≤ 2^{n(H(X)+ε)} for n sufficiently large.
 Pr{A_ε⁽ⁿ⁾} > 1 − ε for n sufficiently large.

Typical Set: A Pictorial Description

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Consider all the instances $(x_1, x_2, ..., x_n) \in \mathcal{X}^n$ of i.i.d. $(X_1, X_2, ..., X_n)$ with distribution p(x).





Source Coding (Data Compression)

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FIGURE 3.2. Source code using the typical set.

Continuous-Time Information Theory

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 Represent each typical sequence with about nH(X) bits.

Source Coding (Data Compression)



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- Represent each non-typical sequence with about n log |X| bits.

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Source Coding (Data Compression)



FIGURE 3.2. Source code using the typical set.

- Represent each typical sequence with about nH(X) bits.
- Represent each non-typical sequence with about n log |X| bits.
- Then we have a one-to-one and easily decodable code.

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Shannon's Source Coding Theorem

$$\mathbb{E}[I(X_1,\ldots,X_n)] = \sum_{x_1,\ldots,x_n} p(x_1,\ldots,x_n)I(x_1,\ldots,x_n)$$

$$\mathbb{E}[I(X_1,...,X_n)] = \sum_{x_1,...,x_n} p(x_1,...,x_n)I(x_1,...,x_n)$$
$$= \sum_{x_1,...,x_n \in A_{\epsilon}^{(n)}} p(x_1,...,x_n)I(x_1,...,x_n) + \sum_{x_1,...,x_n \notin A_{\epsilon}^{(n)}} p(x_1,...,x_n)I(x_1,...,x_n)$$

$$\mathbb{E}[I(X_{1},...,X_{n})] = \sum_{x_{1},...,x_{n}} p(x_{1},...,x_{n})I(x_{1},...,x_{n})$$

$$= \sum_{x_{1},...,x_{n} \in \mathcal{A}_{\epsilon}^{(n)}} p(x_{1},...,x_{n})I(x_{1},...,x_{n}) + \sum_{x_{1},...,x_{n} \notin \mathcal{A}_{\epsilon}^{(n)}} p(x_{1},...,x_{n})I(x_{1},...,x_{n})$$

$$= \sum_{x_{1},...,x_{n} \in \mathcal{A}_{\epsilon}^{(n)}} p(x_{1},...,x_{n})nH(X) + \sum_{x_{1},...,x_{n} \notin \mathcal{A}_{\epsilon}^{(n)}} p(x_{1},...,x_{n})n\log|\mathcal{X}|$$

$$\mathbb{E}[l(X_1,\ldots,X_n)] = \sum_{x_1,\ldots,x_n} p(x_1,\ldots,x_n) l(x_1,\ldots,x_n)$$
$$= \sum_{x_1,\ldots,x_n \in A_{\epsilon}^{(n)}} p(x_1,\ldots,x_n) l(x_1,\ldots,x_n) + \sum_{x_1,\ldots,x_n \notin A_{\epsilon}^{(n)}} p(x_1,\ldots,x_n) l(x_1,\ldots,x_n)$$
$$= \sum_{x_1,\ldots,x_n \in A_{\epsilon}^{(n)}} p(x_1,\ldots,x_n) n H(X) + \sum_{x_1,\ldots,x_n \notin A_{\epsilon}^{(n)}} p(x_1,\ldots,x_n) n \log |\mathcal{X}| \approx n H(X).$$

The average bits needed is

$$\mathbb{E}[I(X_{1},...,X_{n})] = \sum_{x_{1},...,x_{n}} p(x_{1},...,x_{n})I(x_{1},...,x_{n})$$

$$= \sum_{x_{1},...,x_{n} \in \mathcal{A}_{\epsilon}^{(n)}} p(x_{1},...,x_{n})I(x_{1},...,x_{n}) + \sum_{x_{1},...,x_{n} \notin \mathcal{A}_{\epsilon}^{(n)}} p(x_{1},...,x_{n})I(x_{1},...,x_{n})$$

$$= \sum_{x_{1},...,x_{n} \in \mathcal{A}_{\epsilon}^{(n)}} p(x_{1},...,x_{n})nH(X) + \sum_{x_{1},...,x_{n} \notin \mathcal{A}_{\epsilon}^{(n)}} p(x_{1},...,x_{n})n\log|\mathcal{X}| \approx nH(X).$$

Source Coding Theorem

For any information source distributed according to $X_1, X_2, \dots \sim p(x)$, the compression rate is always greater than H(X), but it can be arbitrarily close to H(X).

Communication Channel: Definition

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FIGURE 7.8. Communication channel.

Communication Channel: Definition



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• A message W results in channel inputs $X_1(W), \ldots, X_n(W)$;

Communication Channel: Definition



FIGURE 7.8. Communication channel.

 A message W results in channel inputs X₁(W),..., Xₙ(W);
 And they are received as a random sequence Y₁,..., Yₙ ~ p(y₁,..., yₙ|x₁,..., xₙ).

Communication Channel: Definition



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- And they are received as a random sequence

$$Y_1,\ldots,Y_n\sim p(y_1,\ldots,y_n|x_1,\ldots,x_n).$$

- ► The receiver then guesses the index W by an appropriate decoding rule \$\hat{W} = g(Y_1, ..., Y_n)\$.
- The receiver makes an error if W is not the same as W that was transmitted.

Communication Channel: An Example

Communication Channel: An Example Binary Symmetric Channel

$$p(Y = 0|X = 0) = 1 - p,$$

 $p(Y = 0|X = 1) = p,$

$$p(Y = 1 | X = 0) = p,$$

 $p(Y = 1 | X = 1) = 1 - p.$

Communication Channel: An Example Binary Symmetric Channel

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 $p(Y = 1 | X = 1) = 1 - p.$



FIGURE 7.5. Binary symmetric channel. C = 1 - H(p) bits.

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Tradeoff between Speed and Reliability
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Speed

To transmit 1: we transmit 1. It is likely that we receive 0. Note that the transmission rate is 1.

Tradeoff between Speed and Reliability

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To transmit 1: we transmit 1. It is likely that we receive 0. Note that the transmission rate is 1.

Reliability

To transmit 1: we transmit 11111. Though it is likely that we receive something else, such as 11011, but more likely than not, we can correct the possible error. Note that the transmission rate is however 1/5.

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Shannon's Channel Coding Theorem: Statement

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Channel Coding Theorem

For any discrete memoryless channel, asymptotically perfect transmission rate **below** the **capacity**

$$C = \max_{p(x)} I(X; Y)$$

is always possible, but is not possible **above** the capacity.

Shannon's Channel Coding Theorem: Proof



FIGURE 7.7. Channels after *n* uses.

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Shannon's Channel Coding Theorem: Proof

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- The total number of possible typical Y sequences is approximately 2^{nH(Y)}. This set has to be divided into sets of size 2^{nH(Y|X)} corresponding to the different input X sequences.
- ► The total number of disjoint sets is less than or equal to $2^{n(H(Y)-H(Y|X))} = 2^{nI(X;Y)}$. Hence, we can send at most approximately $2^{nI(X;Y)}$ distinguishable sequences of length *n*.

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Capacity of Binary Symmetric Channels

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The **capacity** of a binary symmetric channel with crossover probability p is C = 1 - H(p), where

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= $H(Y) - \sum_{x} p(x)H(Y|X = x)$

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= $H(Y) - \sum_{x} p(x)H(p)$
= $H(Y) - H(p)$
 $\leq 1 - H(p).$

Continuous-Time Information Theory

Capacity of Additive White Gaussian Channels

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The capacity of an additive white Gaussian channel Y = X + Z, where $\mathbb{E}[X^2] \leq P$ and $Z \sim N(0, 1)$, is $C = \frac{1}{2}\log(1 + P)$. Proof.

I(X;Y) = H(Y) - H(Y|X)



Capacity of Additive White Gaussian Channels

$$I(X;Y) = H(Y) - H(Y|X)$$

= $H(Y) - H(X + Z|X)$

Capacity of Additive White Gaussian Channels

$$I(X; Y) = H(Y) - H(Y|X)$$

= $H(Y) - H(X + Z|X)$
= $H(Y) - H(Z|X)$



Capacity of Additive White Gaussian Channels

$$I(X; Y) = H(Y) - H(Y|X)$$

= $H(Y) - H(X + Z|X)$
= $H(Y) - H(Z|X)$
= $H(Y) - H(Z)$



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$$\begin{split} I(X;Y) &= H(Y) - H(Y|X) \\ &= H(Y) - H(X+Z|X) \\ &= H(Y) - H(Z|X) \\ &= H(Y) - H(Z) \\ &\leq \frac{1}{2} \log 2\pi e (1+P) - \frac{1}{2} \log 2\pi e \end{split}$$

Guangyue Han

Capacity of Additive White Gaussian Channels

$$\begin{split} I(X;Y) &= H(Y) - H(Y|X) \\ &= H(Y) - H(X + Z|X) \\ &= H(Y) - H(Z|X) \\ &= H(Y) - H(Z) \\ &\leq \frac{1}{2} \log 2\pi e (1+P) - \frac{1}{2} \log 2\pi e \\ &= \frac{1}{2} \log (1+P). \end{split}$$

Memoryless Channels

Memoryless Channels

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Memoryless Channels

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- Channel inputs are independent and identically distributed.

Memoryless Channels

- Channel transitions are characterized by time-invariant transition probabilities {p(y|x)}.
- Channel inputs are independent and identically distributed.
- Representative examples include (memoryless) binary symmetric channels and additive white Gaussian channels.

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Capacity of Memoryless Channels

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Capacity of Memoryless Channels



Shannon's channel coding theorem

$$C = \sup_{p(x)} I(X; Y)$$

=
$$\sup_{p(x)} - \sum_{x,y} p(x,y) \log \frac{p(x,y)}{p(x)p(y)}.$$

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Capacity of Memoryless Channels



Shannon's channel coding theorem

$$C = \sup_{p(x)} I(X; Y)$$

=
$$\sup_{p(x)} -\sum_{x,y} p(x,y) \log \frac{p(x,y)}{p(x)p(y)}.$$

The Blahut-Arimoto algorithm (BAA)



Memory Channels

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Channel transitions are characterized by probabilities {p(y_i|x₁,...,x_i, y₁,..., y_{i-1}, s_i)}, where channel outputs are possibly dependent on previous and current channel inputs and previous outputs and current channel state; for example, inter-symbol interference channels, flash memory channels, Gilbert-Elliot channels.

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- Channel inputs may have to satisfy certain constraints which necessitate dependence among channel inputs; for example, (d, k)-RLL constraints, more generally, finite-type constraints.
- Such channels are widely used in a variety of real-life applications, including magnetic and optical recording, solid state drives, communications over band-limited channels with inter-symbol interference.

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Capacity of Memory Channels

Capacity of Memory Channels

Despite a great deal of efforts by Zehavi and Wolf [1988], Mushkin and Bar-David [1989], Shamai and Kofman [1990], Goldsmith and Varaiya [1996], Arnold, Loeliger, Vontobel, Kavcic and Zeng [2006], Holliday, Goldsmith, and Glynn [2006], Vontobel, Kavcic, Arnold and Loeliger [2008], Pfister [2011], Permuter, Asnani and Weissman [2013], Han [2015], ...

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Capacity of Memory Channels

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Capacity of Memory Channels



Shannon's channel coding theorem

$$C = \sup_{p(x)} I(X; Y)$$

= $\sup_{p(x)} \lim_{n \to \infty} -\frac{1}{n} \sum_{x_1^n, y_1^n} p(x_1^n, y_1^n) \log \frac{p(x_1^n, y_1^n)}{p(x_1^n) p(y_1^n)}.$

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The Generalized Blahut-Arimoto algorithm (GBAA) by Vontobel, Kavcic, Arnold and Loeliger [2008]

Algorithm 45 (Generalized BAA): Let Q = Q(B) be a given FSMS manifold and let W be the channel law of a given FSMC. Let $\{Q_{ij}^{(0)}\} \in \operatorname{relint}(Q)$ be some initial (freely chosen) FSMS process. For iterations $r = 0, 1, 2, \ldots$, perform alternatively the following two steps.

- First Step: For each $(i, j) \in \mathcal{B}$ calculate $T_{ij}^{(r)} \triangleq T_{ij}(Q_{ij}^{(r)}, W)$ according to Definition 41. The values $T_{ij}^{(r)}$ can be approximated by the procedure given in Section V-C.
- Second Step: The new FSMS process {Q^(r+1)_{ij}} is chosen to maximize Ψ(Q^(r)_{ij}, Q_{ij}, W), i.e.,

$$\left\{Q_{ij}^{\langle r+1\rangle}\right\} \triangleq \arg\max_{\{Q_{ij}\}\in\mathcal{Q}}\Psi\left(Q_{ij}^{\langle r\rangle},Q_{ij},W\right)$$

and is calculated according to the algorithm in Lemma 44 with inputs $\{\bar{Q}_{ij}\} \triangleq \{Q_{ij}^{(r)}\}$ and W and output $\{Q_{ij}^{(r+1)}\} \triangleq \{Q_{ij}^*\}.$

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Convergence of the GBAA

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Concavity Conjecture [Vontobel *et al.* 2008] I(X; Y) and H(X|Y) are both concave with respect to a chosen parameterization.

Unfortunately, the concavity conjecture is **not** true in general [Li and Han, 2013].

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A Randomized Algorithm [Han 2015]

With appropriately chosen step sizes $a_n = 1/n^a$, a > 0,

$$\theta_{n+1} = \theta_n + a_n g_{n^b}(\theta_n),$$

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With appropriately chosen step sizes $a_n = 1/n^a$, a > 0,

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θ₀ is randomly selected from the parameter space Θ;
g_{n^b}(θ) is a simulator for l'(X(θ); Y(θ));

$$0 < \beta < \alpha < 1/3, \ b > 0, \ 2a + b - 3b\beta > 1,$$

here, α, β are some "hidden" parameters involved in the definition of $g_{n^b}(\theta)$.

Define

$$q = q(n) \triangleq n^{\beta}, \ p = p(n) \triangleq n^{\alpha}, \ k = k(n) \triangleq n/(n^{\alpha} + n^{\beta}).$$

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For any j with $iq + (i - 1)p + 1 \le j \le iq + ip$, define

$$W_j = -\left(\frac{p'(Y_{j-\lfloor q/2 \rfloor})}{p(Y_{j-\lfloor q/2 \rfloor})} + \dots + \frac{p'(Y_j|Y_{j-\lfloor q/2 \rfloor}^{j-1})}{p(Y_j|Y_{j-\lfloor q/2 \rfloor}^{j-1})}\right)\log p(Y_j|Y_{j-\lfloor q/2 \rfloor}^{j-1}),$$

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and furthermore

$$\zeta_i \triangleq W_{iq+(i-1)p+1} + \cdots + W_{iq+ip}, \quad S_n \triangleq \sum_{i=1}^{k(n)} \zeta_i.$$

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Our simulator for I'(X; Y):

$$g_n(X_1^n, Y_1^n) = H'(X_2|X_1) + S_n(Y_1^n)/(kp) - S_n(X_1^n, Y_1^n)/(kp).$$

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Convergence of Our Algorithm

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Convergence and convergence rate with concavity

If I(X; Y) is concave with respect to θ , then θ_n converges to the unique capacity achieving distribution θ^* almost surely. And for any τ with $2a + b - 3b\beta - 2\tau > 1$, we have

$$|\theta_n - \theta^*| = \tilde{O}(n^{-\tau}).$$

Continuous-Time Information Theory

The Ideas for the Proofs

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Analyticity result [Han, Marcus, 2006]

The entropy rate of hidden Markov chains is analytic.

Refinements of the Shannon-MaMillan-Breiman theorem [Han, 2012] Limit theorems for the sample entropy of hidden Markov chains hold.

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The refinement results confirm that using Monte Carlo simulations, I(X; Y) and its derivatives can be "well-approximated".

Continuous-Time Information Theory

Continuous-Time Gaussian Non-Feedback Channels

Consider the following continuous-time Gaussian channel:

$$Y(t)=\sqrt{\mathit{snr}}\int_0^t X(s)\mathit{ds}+B(t),\ t\in[0,\,T],$$

where $\{B(t)\}$ is the standard Brownian motion.

Continuous-Time Gaussian Non-Feedback Channels

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Theorem (Ducan 1970) The following I-CMMSE relationship holds:

$$I(X_0^T; Y_0^T) = \frac{1}{2} \mathbb{E} \int_0^T (X(s) - \mathbb{E}[X(s)|Y_0^s])^2 ds.$$

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Theorem (Guo, Shamai and Verdu 2005) The following I-MMSE relationship holds:

$$\frac{d}{dsnr}I(X_0^{\mathcal{T}};Y_0^{\mathcal{T}}) = \frac{1}{2}\mathbb{E}\int_0^{\mathcal{T}}(X(s) - \mathbb{E}[X(s)|Y_0^{\mathcal{T}}])^2 ds.$$

Continuous-Time Gaussian Feedback Channels

Consider the following continuous-time Gaussian feedback channel:

$$Y(t) = \sqrt{\operatorname{snr}} \int_0^t X(s, M, Y_0^s) ds + B(t), \ t \in [0, T],$$

where $\{B(t)\}$ is the standard Brownian motion.

Continuous-Time Gaussian Feedback Channels

Continuous-Time Gaussian Feedback Channels Theorem (Kadota, Zakai and Ziv 1971) The following I-CMMSE relationship:

$$I(M; Y_0^T) = \frac{1}{2} \mathbb{E} \int_0^T (X(s, M, Y_0^s) - \mathbb{E}[X(s, M, Y_0^s) | Y_0^s])^2 \, ds.$$

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Theorem (Han and Song 2016) The following I-MMSE relationship holds:

$$\frac{d}{dsnr}I(M;Y_0^T) = \frac{1}{2}\int_0^T \mathbb{E}\left[\left(X(s) - \mathbb{E}[X(s)|Y_0^T]\right)^2\right]ds$$
$$+snr\int_0^T \mathbb{E}\left[\left(X(s) - \mathbb{E}\left[X(s)|Y_0^T\right]\right)\frac{d}{dsnr}X(s)\right]ds.$$
Memory Channels

Continuous-Time Information Theory 0000000

Capacity of Continuous-Time Gaussian Channels

Capacity of Continuous-Time Gaussian Channels

For either the following continuous-time Gaussian channel:

$$Y(t)=\sqrt{snr}\int_0^t X(s)ds+B(t),\ t\in[0,T],$$

or the following continuous-time Gaussian feedback channel:

$$Y(t) = \sqrt{\operatorname{snr}} \int_0^t X(s, M, Y_0^s) ds + B(t), \ t \in [0, T],$$

the capacity is P/2.

Thank you!