## Universally Decodable

 Matrices for Distributed
## Matrix-Vecto Mutiplication

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## Motivation

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Given:

- a matrix A of size $m \times n$ over the reals;
- a vector x of length $n$ over the reals.

Task: compute the vector $y$ of length $m$ over the reals, where

$$
\mathbf{y} \triangleq \mathbf{A} \cdot \mathbf{x} .
$$

Explicitly:

$$
\left(\begin{array}{c}
y_{1} \\
\vdots \\
y_{m}
\end{array}\right) \triangleq\left(\begin{array}{ccc}
a_{1,1} & \cdots & a_{1, n} \\
\vdots & \cdots & \vdots \\
a_{m, 1} & \cdots & a_{m, n}
\end{array}\right) \cdot\left(\begin{array}{c}
x_{1} \\
\vdots \\
x_{n}
\end{array}\right) .
$$

## Motivation

We can split up the task into several submatrix-vector-multiplication tasks:

| $\mathbf{y}_{0,0}$ |
| :--- | :--- |
| $\mathbf{y}_{0,1}$ |
| $\mathbf{y}_{1,0}$ |
| $\mathbf{y}_{1,1}$ |
| $\mathbf{y}_{2,0}$ |
| $\mathbf{y}_{2,1}$ | $\mathbf{A}_{0,0}$

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Idea:

- Use coding theory to alleviate delay issues because of stragglers.


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- Use coding theory to alleviate delay issues because of stragglers.
- Unavailable partial results can be seen as erasures.


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Idea:

- Coding scheme should take advantage of the fact that erasures are correlated.

Erasures are correlated because
if a partial result by one of the workers is not available, then all subsequent results by the same worker are not available either.

## Motivation

We can split up the task into several submatrix-vector-multiplication tasks:


Idea:

- Base coding scheme on so-called universally decodable matrices (UDMs).


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We can split up the task into several submatrix-vector-multiplication tasks:


Idea:

- Base coding scheme on so-called universally decodable matrices (UDMs).
- Use companion matrices in order to reduce issues with condition numbers when adapting a coding scheme over some finite field to a coding scheme over the reals.


## Motivation

Bad condition number of unsuitably chosen encoding matrices is an issue.


## Context (Part 1/2)

- Q. Yu, M. Maddah-Ali, and S. Avestimehr, "Polynomial codes: an optimal design for high-dimensional coded matrix multiplication," in Proc. of Adv. in Neural Inf. Proc. Sys. (NIPS), 2017, pp. 4403-4413.
- L. Tang, K. Konstantinidis, and A. Ramamoorthy, "Erasure coding for distributed matrix multiplication for matrices with bounded entries," IEEE Comm. Lett., vol. 23, no. 1, pp. 8-11, 2019.
- K. Lee, C. Suh, and K. Ramchandran, "High-dimensional coded matrix multiplication," in IEEE Int. Symp. Inf. Theory, 2017, pp. 2418-2422.
- K. Lee, M. Lam, R. Pedarsani, D. Papailiopoulos, and K. Ramchandran, "Speeding up distributed machine learning using codes," IEEE Trans. Inf. Theory, vol. 64, no. 3, pp. 1514-1529, 2018.
- S. Dutta, V. Cadambe, and P. Grover, "Short-dot: Computing large linear transforms distributedly using coded short dot products," in Proc. of Adv. in Neural Inf. Proc. Sys. (NIPS), 2016, pp. 2100-2108.


## Context (Part $/ 2$ )

- A. Mallick, M. Chaudhari, and G. Joshi, "Rateless codes for near-perfect load balancing in distributed matrix-vector multiplication," preprint, 2018. arXiv: 1804.10331.
- S. Wang, J. Liu, and N. B. Shroff, "Coded sparse matrix multiplication," in Proc. 35th Int. Conf. Mach. Learning, ICML, 2018, pp. 5139-5147.
- S. Kiani, N. Ferdinand, and S. C. Draper, "Exploitation of stragglers in coded computation," in IEEE Int. Symp. Inf. Theory, 2018, pp. 1988-1992.
- A. B. Das, L. Tang, and A. Ramamoorthy, "C³LES: Codes for coded computation that leverage stragglers," in IEEE Inf. Th. Workshop, 2018, pp. 1-5.
- N. Raviv, Y. Cassuto, R. Cohen, and M. Schwartz, "Erasure correction of scalar codes in the presence of stragglers," in IEEE Int. Symp. Inf. Theory, 2018, pp. 1983-1987.
- N. Raviv, Q. Yu, J. Bruck, and S. Avestimehr, "Download and access tradeoffs in Lagrange coded computing," in IEEE Int. Symp. Inf. Theory, 2019.


## Overview

## Overview

- Motivation
- A communication system with $L$ parallel channels
$\Rightarrow$ Coding for this system using universally decodable matrices
- Embedding into the reals
$\Rightarrow$ Companion matrices


## For more details:

- A. Ramamoorthy, L. Tang, and P. O. Vontobel, "Universally decodable matrices for distributed matrix-vector multiplication," Proc. IEEE Int. Symp. Inf. Theory, Paris, France, pp. 1777-1781, July 2019.
- arXiv: 1901.10674


## Communication system

## with $L$ parallel channels

## Comm. System with $L$ Parallel Channels



## Comm. System with $L$ Parallel Channels



$$
\begin{aligned}
&\left(\begin{array}{lll}
u_{0} & \cdots & u_{n-1}
\end{array}\right) \mapsto\left(\begin{array}{ccc}
x_{0,0} & \cdots & x_{0, n-1} \\
\vdots & \vdots & \vdots \\
x_{L-1,0} & \cdots & x_{L-1, n-1}
\end{array}\right) \\
& \Rightarrow\left(\begin{array}{ccc}
y_{0,0} & \cdots & y_{0, n-1} \\
\vdots & \vdots & \vdots \\
y_{L-1,0} & \cdots & y_{L-1, n-1}
\end{array}\right) \Rightarrow\left(\begin{array}{lll}
\hat{u}_{0} & \cdots & \hat{u}_{n-1}
\end{array}\right)
\end{aligned}
$$

## Comm. System with $L$ Parallel Channels


E.g. $L=4, n=3$.

$$
\left(\begin{array}{lll}
u_{0} & u_{1} & u_{2}
\end{array}\right) \mapsto\left(\begin{array}{ccc}
x_{0,0} & x_{0,1} & x_{0,2} \\
x_{1,0} & x_{1,1} & x_{1,2} \\
x_{2,0} & x_{2,1} & x_{2,2} \\
x_{3,0} & x_{3,1} & x_{3,2}
\end{array}\right) \Rightarrow\left(\begin{array}{lll}
y_{0,0} & y_{0,1} & y_{0,2} \\
y_{1,0} & y_{1,1} & y_{1,2} \\
y_{2,0} & y_{2,1} & y_{2,2} \\
y_{3,0} & y_{3,1} & y_{3,2}
\end{array}\right) \Rightarrow\left(\begin{array}{lll}
\hat{u}_{0} & \hat{u}_{1} & \hat{u}_{2}
\end{array}\right)
$$

## Comm. System with $L$ Parallel Channels


E.g. $L=4, n=3, q=3$.

$$
\left(\begin{array}{lll}
1 & 1 & 2
\end{array}\right) \mapsto\left(\begin{array}{lll}
1 & 1 & 2 \\
2 & 1 & 1 \\
1 & 2 & 2 \\
2 & 0 & 2
\end{array}\right) \Rightarrow\left(\begin{array}{ccc}
? & ? & ? \\
2 & ? & ? \\
? & ? & ? \\
2 & 0 & ?
\end{array}\right) \Rightarrow\left(\begin{array}{lll}
\hat{u}_{0} & \hat{u}_{1} & \hat{u}_{2}
\end{array}\right)
$$

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2 & 1 & 1 \\
1 & 2 & 2 \\
2 & 0 & 2
\end{array}\right) \Rightarrow\left(\begin{array}{ccc}
? & ? & ? \\
2 & ? & ? \\
? & ? & ? \\
2 & 0 & ?
\end{array}\right) \Rightarrow\left(\begin{array}{lll}
\hat{u}_{0} & \hat{u}_{1} & \hat{u}_{2}
\end{array}\right)
$$

The channels are such that if $y_{\ell, t}$ is erased then also $y_{\ell, t^{\prime}}$ is erased for all $t^{\prime}>t$.

## Comm. System with $L$ Parallel Channels


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$$
\left.\left(\begin{array}{lll}
1 & 1 & 2
\end{array}\right) \mapsto\left(\begin{array}{lll}
1 & 1 & 2 \\
2 & 1 & 1 \\
1 & 2 & 2 \\
2 & 0 & 2
\end{array}\right) \Rightarrow\left(\begin{array}{ccc}
? & ? & ? \\
2 & ? & ? \\
? & ? & ? \\
2 & 0 & ?
\end{array}\right) k_{0}=0 . k_{1}=1 . k_{2}=0.1 \begin{array}{lll}
k_{3}=2
\end{array} \hat{u}_{1} \quad \hat{u}_{2}\right)
$$

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1 & 1 & 2 \\
2 & 1 & 1 \\
1 & 2 & 2 \\
2 & 0 & 2
\end{array}\right) \Rightarrow\left(\begin{array}{ccc}
? & ? & ? \\
2 & ? & ? \\
? & ? & ? \\
2 & 0 & ?
\end{array}\right) \begin{aligned}
& k_{0}=0 \\
& k_{1}=1 \\
& k_{2}=0 \\
& k_{3}=2
\end{aligned} \Rightarrow\left(\begin{array}{lll}
\hat{u}_{0} & \hat{u}_{1} & \hat{u}_{2}
\end{array}\right)
$$

We want unique decodability as long as $\sum_{\ell \in[L]} k_{\ell} \geq n$,

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1 & 1 & 2 \\
2 & 1 & 1 \\
1 & 2 & 2 \\
2 & 0 & 2
\end{array}\right) \Rightarrow\left(\begin{array}{ccc}
? & ? & ? \\
2 & ? & ?
\end{array}\right) \begin{array}{l}
k_{0}=0 \\
k_{1}=1 \\
?
\end{array} \begin{array}{l}
? \\
2
\end{array} 0 \begin{array}{l}
?
\end{array}\right) k_{2}=0 \quad\left(\begin{array}{lll}
k_{3}=2
\end{array} \hat{u}_{1} \quad \hat{u}_{2}\right)
$$

We want unique decodability as long as $\sum_{\ell \in[L]} k_{\ell} \geq n$, here: $k_{0}+k_{1}+k_{2}+k_{3} \geq 3$.

## Comm. System with $L$ Parallel Channels



## Comm. System with $L$ Parallel Channels



For reasons of simplicity, we would like the encoding to be linear:

$$
\mathbf{x}_{0}=\mathbf{u} \cdot \mathbf{G}_{0}, \quad \ldots, \quad \mathbf{x}_{L-1}=\mathbf{u} \cdot \mathbf{G}_{L-1}
$$

where $\mathbf{G}_{0}, \ldots, \mathbf{G}_{L-1}$ are $n \times n$ matrices.

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$$

where $\mathbf{G}_{0}, \ldots, \mathbf{G}_{L-1}$ are $n \times n$ matrices.
Definition: If the above matrices lead to unique decodability for any $k_{0}, \ldots, k_{L-1}$ with $\sum_{\ell \in[L]} k_{\ell} \geq n$, then we call these matrices universally decodable matrices (UDMs).

## Comm. System with $L$ Parallel Channels


E.g. $L=2, n=5$, any $q$. The matrices $\mathrm{G}_{0}$ and $\mathrm{G}_{1}$ are UDMs:

$$
\mathbf{G}_{0}=\left(\begin{array}{lllll}
1 & 0 & 0 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 0 & 0 & 1
\end{array}\right), \mathbf{G}_{1}=\left(\begin{array}{lllll}
0 & 0 & 0 & 0 & 1 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
1 & 0 & 0 & 0 & 0
\end{array}\right) .
$$

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1 & 0 & 0 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 0 & 0 & 1
\end{array}\right), \mathbf{G}_{1}=\left(\begin{array}{lllll}
0 & 0 & 0 & 0 & 1 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
1 & 0 & 0 & 0 & 0
\end{array}\right) .
$$

$$
\mathbf{u}=\left(\begin{array}{lllll}
u_{0} & u_{1} & u_{2} & u_{3} & u_{4}
\end{array}\right)
$$

## Comm. System with $L$ Parallel Channels


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$$
\left.\begin{array}{rl}
\mathbf{G}_{0}=\left(\begin{array}{ccccc}
1 & 0 & 0 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 0 & 0 & 1
\end{array}\right), \mathbf{G}_{1}=\left(\begin{array}{ccccc}
0 & 0 & 0 & 0 & 1 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
1 & 0 & 0 & 0 & 0
\end{array}\right) . \\
\mathbf{u}=\left(\begin{array}{llll}
u_{0} & u_{1} & u_{2} & u_{3}
\end{array} u_{4}\right.
\end{array}\right), \quad \mathbf{x}_{0}=\left(\begin{array}{lllll}
u_{0} & u_{1} & u_{2} & u_{3} & u_{4}
\end{array}\right), .
$$

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$$
\left.\begin{array}{c}
\mathbf{G}_{0}=\left(\begin{array}{ccccc}
1 & 0 & 0 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 0 & 0 & 1
\end{array}\right), \mathbf{G}_{1}=\left(\begin{array}{ccccc}
0 & 0 & 0 & 0 & 1 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
1 & 0 & 0 & 0 & 0
\end{array}\right) . \\
\mathbf{u}=\left(u_{0} u_{1} u_{2} u_{3} u_{4}\right), \mathbf{x}_{0}=\left(\begin{array}{llll}
u_{0} & u_{1} & u_{2} & u_{3}
\end{array} u_{4}\right.
\end{array}\right), \mathbf{x}_{1}=\left(\begin{array}{llll}
u_{4} & u_{3} & u_{2} & u_{1} \\
u_{0}
\end{array}\right) .
$$

## Comm. System with $L$ Parallel Channels


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1 & 0 & 0 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 0 & 0 & 1
\end{array}\right), \mathbf{G}_{1}=\left(\begin{array}{lllll}
0 & 0 & 0 & 0 & 1 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
1 & 0 & 0 & 0 & 0
\end{array}\right) .
$$

## Comm. System with $L$ Parallel Channels


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$$
\mathbf{G}_{0}=\left(\begin{array}{lllll}
1 & 0 & 0 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 0 & 0 & 1
\end{array}\right), \mathbf{G}_{1}=\left(\begin{array}{lllll}
0 & 0 & 0 & 0 & 1 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
1 & 0 & 0 & 0 & 0
\end{array}\right) .
$$

$$
\mathbf{u}=\left(\begin{array}{lllll}
u_{0} & u_{1} & u_{2} & u_{3} & u_{4}
\end{array}\right)
$$

## Comm. System with $L$ Parallel Channels


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\mathbf{G}_{0}=\left(\begin{array}{ccccc}
1 & 0 & 0 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 0 & 0 & 1
\end{array}\right), \mathbf{G}_{1}=\left(\begin{array}{ccccc}
0 & 0 & 0 & 0 & 1 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
1 & 0 & 0 & 0 & 0
\end{array}\right) . \\
\mathbf{u}=\left(\begin{array}{llll}
u_{0} & u_{1} & u_{2} & u_{3}
\end{array} u_{4}\right.
\end{array}\right), \quad \mathbf{y}_{0} \quad=\left(\begin{array}{llll}
u_{0} & u_{1} & u_{2} & ?
\end{array}\right), .
$$

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E.g. $L=2, n=5$, any $q$. The matrices $\mathrm{G}_{0}$ and $\mathrm{G}_{1}$ are UDMs:

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\mathbf{G}_{0}=\left(\begin{array}{ccccc}
1 & 0 & 0 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 0 & 0 & 1
\end{array}\right), \mathbf{G}_{1}=\left(\begin{array}{ccccc}
0 & 0 & 0 & 0 & 1 \\
0 & 0 & 0 & 1 & 0 \\
0 & 0 & 1 & 0 & 0 \\
0 & 1 & 0 & 0 & 0 \\
1 & 0 & 0 & 0 & 0
\end{array}\right) . \\
\mathbf{u}=\left(\begin{array}{llll}
u_{0} & u_{1} & u_{2} & u_{3}
\end{array} u_{4}\right.
\end{array}\right), \quad \mathbf{y}_{0}=\left(\begin{array}{llll}
u_{0} & u_{1} & u_{2} & ?
\end{array}\right), \mathbf{x}_{1}=\left(\begin{array}{lll}
u_{4} & u_{3} & ?
\end{array}\right) ? .\right\}
$$

## Comm. System with $L$ Parallel Channels


E.g. $L=4, n=3, q=3$. The matrices $\mathbf{G}_{0}, \mathbf{G}_{1}, \mathbf{G}_{2}, \mathbf{G}_{3}$ are UDMs:

$$
\mathbf{G}_{0}=\left(\begin{array}{lll}
1 & 0 & 0 \\
0 & 1 & 0 \\
0 & 0 & 1
\end{array}\right), \quad \mathbf{G}_{1}=\left(\begin{array}{lll}
0 & 0 & 1 \\
0 & 1 & 0 \\
1 & 0 & 0
\end{array}\right), \quad \mathbf{G}_{2}=\left(\begin{array}{lll}
1 & 0 & 0 \\
1 & 1 & 0 \\
1 & 2 & 1
\end{array}\right), \quad \mathbf{G}_{3}=\left(\begin{array}{lll}
1 & 0 & 0 \\
2 & 1 & 0 \\
1 & 1 & 1
\end{array}\right) .
$$

## Comm. System with $L$ Parallel Channels

What does unique decodability imply for the matrices $\mathrm{G}_{0}, \ldots, \mathrm{G}_{L-1}$ ?

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What does unique decodability imply for the matrices $\mathrm{G}_{0}, \ldots, \mathrm{G}_{L-1}$ ?


For any $k_{0}, \ldots, k_{L-1}$ with $\sum_{\ell \in[L]} k_{\ell} \geq n$ the matrix $\mathbf{G}$ must have full rank.

## Comm. System with $L$ Parallel Channels

- Another motivation for this channel model: paper by Tavildar and Viswanath, "Approximately universal codes over slow fading channels", IEEE Trans. Inf. Theory, IT-52, no. 7, pp. 3233-3258, July 2006.


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- Another motivation for this channel model: paper by Tavildar and Viswanath, "Approximately universal codes over slow fading channels", IEEE Trans. Inf. Theory, IT-52, no. 7, pp. 3233-3258, July 2006.
- Consider slow-fading (point-to-point) MIMO channel

$$
\mathbf{y}[m]=\mathbf{H} \cdot \mathbf{x}[m]+\mathbf{w}[m] .
$$

The complex matrix of fading gains H stays constant over the time-scale of communication; we suppose the exact characterization of $H$ is known to the receiver while the transmitter has only access to its statistical characterization.

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- The focus in the paper is on the high-SNR regime.


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The complex matrix of fading gains H stays constant over the time-scale of communication; we suppose the exact characterization of H is known to the receiver while the transmitter has only access to its statistical characterization.

- The focus in the paper is on the high-SNR regime.
- Coding for this channel can be seen as space-time coding.


## Comm. System with $L$ Parallel Channels

- Depending on what $h_{\ell}$ is, we can recover more or fewer of the most-significant bits.


## Comm. System with $L$ Parallel Channels

- Depending on what $h_{\ell}$ is, we can recover more or fewer of the most-significant bits.
- Assume $L=2$ : channel is not in outage if

$$
\log \left(1+\left|h_{0}\right|^{2} \mathrm{SNR}\right)+\log \left(1+\left|h_{1}\right|^{2} \mathrm{SNR}\right)>2 R
$$

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$$

- Assume that $h_{0}$ and $h_{1}$ are such that

$$
\log \left(1+\left|h_{0}\right|^{2} \mathrm{SNR}\right)>2 k_{0}, \quad \log \left(1+\left|h_{1}\right|^{2} \mathrm{SNR}\right)>2 k_{1} .
$$

for some $k_{0}$ and $k_{1}$, i.e. we can recover $k_{0}$ bits from the zeroth channel and $k_{1}$ bits from the first channel.

## Comm. System with $L$ Parallel Channels

- Depending on what $h_{\ell}$ is, we can recover more or fewer of the most-significant bits.
- Assume $L=2$ : channel is not in outage if

$$
\log \left(1+\left|h_{0}\right|^{2} \mathrm{SNR}\right)+\log \left(1+\left|h_{1}\right|^{2} \mathrm{SNR}\right)>2 R
$$

- Assume that $h_{0}$ and $h_{1}$ are such that

$$
\log \left(1+\left|h_{0}\right|^{2} \mathrm{SNR}\right)>2 k_{0}, \quad \log \left(1+\left|h_{1}\right|^{2} \mathrm{SNR}\right)>2 k_{1} .
$$

for some $k_{0}$ and $k_{1}$, i.e. we can recover $k_{0}$ bits from the zeroth channel and $k_{1}$ bits from the first channel.

- Not being in outage means that $k_{0}+k_{1} \geq R$.


## Coding via Evaluation

## Coding via Evaluation (first setup)

Encoding map (evaluation map):

$$
\left(u_{0}, u_{1}, u_{2}\right) \mapsto\left(f\left(\beta_{0}\right), f\left(\beta_{1}\right), f\left(\beta_{2}\right), f\left(\beta_{3}\right), f\left(\beta_{4}\right), f\left(\beta_{5}\right), f\left(\beta_{6}\right), f\left(\beta_{7}\right), f\left(\beta_{8}\right)\right),
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where $f(x)=u_{0} x^{0}+u_{1} x^{1}+u_{2} x^{2}$.

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The fundamental theorem of algebra implies that $\operatorname{deg}(f(x)) \geq 3$. However, no quadratic function can have more than two zeros.

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\begin{aligned}
f(x) & =0, \\
\Rightarrow\left(u_{0}, u_{1}, u_{2}\right) & =(0,0,0) .
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$$

where $f(x)=u_{0} x^{0}+u_{1} x^{1}+u_{2} x^{2}$.
Note: the codes that result from this evaluation map are the well-known
Reed-Solomon codes.

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where $f(x)=u_{0} x^{0}+u_{1} x^{1}+u_{2} x^{2}$.

A way to find $\left(u_{0}, u_{1}, u_{2}\right)$ is to specify at least three function values.

## Coding via Evaluation (Second Setup)

However, there are also other quantities that we can specify so that we can find out $\left(u_{0}, u_{1}, u_{2}\right)$.

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For example, knowing

- the function value plus the value of the function derivative for one place and
- the function value at another place,
is sufficient to find $\left(u_{0}, u_{1}, u_{2}\right)$.


## Coding Eictaraid (Second Setup)

However, there are also other quantities that we can specify so that we can find out $\left(u_{0}, u_{1}, u_{2}\right)$.



Consider the following new evaluation map:

$$
\left(\begin{array}{lll}
u_{0} & u_{1} & u_{2}
\end{array}\right) \mapsto\left(\begin{array}{cc}
f\left(\beta_{0}\right) & f^{\prime}\left(\beta_{0}\right) \\
\vdots & \vdots \\
f\left(\beta_{8}\right) & f^{\prime}\left(\beta_{8}\right)
\end{array}\right)
$$

where

$$
f(x)=u_{0} x^{0}+u_{1} x^{1}+u_{2} x^{2} \quad \text { and } \quad f^{\prime}(x)=u_{1} x^{0}+2 u_{2} x^{1}
$$

## Coding via Evaluation (Second Setup)

However, there are also other quantities that we can specify so that we can find out $\left(u_{0}, u_{1}, u_{2}\right)$.



General formula for the evaluation map:

$$
\left(\begin{array}{lll}
u_{0} & \cdots & u_{n-1}
\end{array}\right) \mapsto\left(\begin{array}{cccc}
f^{(0)}\left(\beta_{0}\right) & f^{(1)}\left(\beta_{0}\right) & \cdots & f^{(n-1)}\left(\beta_{0}\right) \\
\vdots & \vdots & \vdots & \vdots \\
f^{(0)}\left(\beta_{L-1}\right) & f^{(1)}\left(\beta_{L-1}\right) & \cdots & f^{(n-1)}\left(\beta_{L-1}\right)
\end{array}\right)
$$

where

$$
f^{(i)}(x)=\sum_{t=0}^{n-1} \frac{t!}{(t-i)!} u_{t} x^{t} \quad \text { for } 0 \leq i \leq n-1
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There is a problem if we want to use this approach when we work over finite fields: if $p$ is the characteristic of $\mathbb{F}_{q}$ then the $i$-th formal derivative is zero for $i \geq p$ and the corresponding channel symbols do not carry any information.

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There is a problem if we want to use this approach when we work over finite fields: if $p$ is the characteristic of $\mathbb{F}_{q}$ then the $i$-th formal derivative is zero for $i \geq p$ and the corresponding channel symbols do not carry any information. However, replacing the formal derivative by the Hasse derivative, this approach works!

## 

General formula for the evaluation map:

$$
\left(\begin{array}{lll}
u_{0} & u_{1} & u_{2}
\end{array}\right) \mapsto\left(\begin{array}{cccc}
\tilde{f}^{(0)}\left(\beta_{0}\right) & \tilde{f}^{(1)}\left(\beta_{0}\right) & \cdots & \tilde{f}^{(n-1)}\left(\beta_{0}\right) \\
\vdots & \vdots & \vdots & \vdots \\
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\end{array}\right)
$$

where we used the Hasse derivatives

$$
\tilde{f}^{(i)}(x)=\sum_{t=0}^{n-1}\binom{t}{i} u_{t} x^{t}=\sum_{t=0}^{n-1} \frac{t!}{i!(t-i)!} u_{t} x^{t} \quad \text { for } 0 \leq i \leq n-1 .
$$

## Coding via Evaluation (second Setup)

Assume that we only receive

- the function value and the derivative for $x=\beta_{2}$ and
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- Because the above mapping is linear it is sufficient to show that the kernel is trivial.


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Case 2:
$f(x) \neq 0$ with at least three zeros (counting with multiplicities).

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## Coding via Evaluation (second setup)

Note that this second interpolation setup is not simply a special case of the first interpolation setup:


Knowing three points where a parabola goes through is sufficient to find out the parameters of the parabola.


Knowing e.g. the derivatives at three points of a parabola is not sufficient to find out the parameters of the parabola.

## Universally decodable matrices (UDMs)

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## Proposition

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## Universally Decodable Matrices

## Proposition

- Let $n$ be some positive integer, let $q$ be some prime power.
- Let $\alpha$ be a primitive element in $\mathbb{F}_{q}$.
(I.e. $\alpha$ is an $(q-1)$-th primitive root of unity.)
- If $L \leq q+1$ then the following $L$ matrices over $\mathbb{F}_{q}$ of size $n \times n$ are (L, $n, q$ )-UDMs:

$$
\mathbf{G}_{0} \triangleq \mathbf{I}_{n}, \quad \mathbf{G}_{1} \triangleq \mathbf{J}_{n}, \quad \mathbf{G}_{2}, \quad \ldots, \quad \mathbf{G}_{L-1}
$$

where

- $\mathbf{J}_{n}$ is an $n \times n$ matrix with ones in the anti-diagonal and zeros otherwise;
- $\left[\mathbf{G}_{\ell+2}\right]_{t, i} \triangleq\binom{t}{i} \alpha^{\ell(t-i)},(\ell, t, i) \in[L-2] \times[n] \times[n]$.


## Universally Decodable Matrices

E.g. $L=4, n=3, q=3$.

$$
\begin{array}{ll}
\mathbf{G}_{0}=\left(\begin{array}{lll}
1 & 0 & 0 \\
0 & 1 & 0 \\
0 & 0 & 1
\end{array}\right), & \mathbf{G}_{1}=\left(\begin{array}{lll}
0 & 0 & 1 \\
0 & 1 & 0 \\
1 & 0 & 0
\end{array}\right), \\
\mathbf{G}_{2}=\left(\begin{array}{lll}
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1 & 1 & 0 \\
1 & 2 & 1
\end{array}\right), & \mathbf{G}_{3}=\left(\begin{array}{lll}
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0 & 0 & 1 \\
0 & 1 & 0 \\
1 & 0 & 0
\end{array}\right), \\
\mathbf{G}_{2}=\left(\begin{array}{lll}
\mathbf{1} & 0 & 0 \\
\mathbf{1} & \mathbf{1} & 0 \\
\mathbf{1} & \mathbf{2} & \mathbf{1}
\end{array}\right), & \mathbf{G}_{3}=\left(\begin{array}{lll}
1 & 0 & 0 \\
2 & 1 & 0 \\
1 & 1 & 1
\end{array}\right) .
\end{array}
$$

Note that $\left[\mathbf{G}_{2}\right]_{t, i} \xlongequal{\triangleq}\binom{t}{i}$, therefore Pascal's triangle plays an important role when constructing these matrices.

## Comments

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- In the last ten years, the resulting codes have also appeared under the name "multiplicity codes" in the theoretical computer science literature.
- The mathematics that is needed is very similar to the mathematics that is needed when studying so-called repeated-root cyclic codes [Castagnoli et al., 1991].
- Are there other constructions of UDMs that are not simply reformulations of the above UDMs? Note that one can show that the given construction is in a certain sense a unique extension of Reed-Solomon codes [Vontobel and Ganesan, 2006].


## Efficient Decoding

- Decoding means that we have to solve the system of linear equations

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- However, decoding is obviously related to finding an interpolation polynomial: the problem at hand can be solved with a variant of Newton's interpolation algorithm. This results in a decoding complexity of $O\left(n^{2}\right)$.


## Generalizations (Part 1/2)

- Remember the encoding that we are using

$$
\mathbf{u} \in \mathbb{F}_{q}^{n} \quad \mapsto \quad \mathbf{x}_{\ell} \in \mathbb{F}_{q}^{n}, \ell \in[L] .
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$$

$\Rightarrow$ The above construction of UDMs can be extended straightforwardly to this new setup.

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- Riemann-Roch theorem gives new proof.


## Generalizations (Part /2/2)

- Remember that for any $k_{0}, \ldots, k_{L-1}$ with $\sum_{k \in[L]} k_{\ell} \geq n$ we required that we can decode uniquely.
- Generalization: for any $k_{0}, \ldots, k_{L-1}$ with $\sum_{k \in[L]} k_{\ell} \geq n+g$ we require that we can decode uniquely for some $g \geq 0$.
$\Rightarrow$ In the same way as Goppa codes / algebraic-geometry codes are generalizations of Reed-Solomon codes, one can construct UDMs that are generalizations of the above UDMs.
- Riemann-Roch theorem gives new proof.
- Hasse-Weil-Serre bound can be used to give new necessary conditions for $L$.


## Back to the setup of interest

## Motivation

We can split up the task into several submatrix-vector-multiplication tasks:


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Idea:

- Coding scheme should take advantage of the fact that erasures are correlated.

Erasures are correlated because
if a partial result by one of the workers is not available, then all subsequent results by the same worker are not available either.

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We can split up the task into several submatrix-vector-multiplication tasks:


Idea:

- Base coding scheme on so-called universally decodable matrices (UDMs).


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We can split up the task into several submatrix-vector-multiplication tasks:


Idea:

- Base coding scheme on so-called universally decodable matrices (UDMs).
- Use companion matrices in order to reduce issues with condition numbers when adapting a coding scheme over some finite field to a coding scheme over the reals.


## Embedding into the reals: companion matrices

## Companion Matrices

Assume that the field $\left\langle\mathbb{F}_{p^{s}},+, \cdot\right\rangle$ is constructed based on the primitive polynomial

$$
\pi(\mathrm{X})=\mathrm{X}^{s}+\pi_{s-1} \mathrm{X}^{s-1}+\cdots+\pi_{1} \mathrm{X}+\pi_{0} \in \mathbb{F}_{p}[\mathrm{X}] .
$$

The companion matrix associated with $\pi(\mathrm{X})$ is defined to be the following matrix of size $s \times s$ over $\mathbb{F}_{p}$ :

$$
\mathbf{C} \xlongequal{\wedge}\left(\begin{array}{ccccc}
0 & 0 & \cdots & 0 & -\pi_{0} \\
1 & 0 & \cdots & 0 & -\pi_{1} \\
0 & 1 & \cdots & 0 & -\pi_{2} \\
\vdots & \vdots & \ddots & \vdots & \vdots \\
0 & 0 & \cdots & 1 & -\pi_{s-1}
\end{array}\right) .
$$

This matrix yields the following field isomorphism:

$$
\left\langle\mathbb{F}_{p^{s}},+, \cdot\right\rangle \cong\left\langle\left\{\mathbf{0}, \mathbf{C}, \mathbf{C}^{2}, \mathbf{C}^{3}, \ldots, \mathbf{C}^{p^{s}-1}\right\},+, \cdot\right\rangle .
$$

## Companion Matrices

Lemma: let $\mathbf{M}$ be a square matrix with entries in $\mathbb{Z}$.
If $M$ satisfies

$$
\operatorname{det}(\mathbf{M}) \neq 0 \quad(\bmod p)
$$

then also

$$
\operatorname{det}(\mathbf{M}) \neq 0 \quad \text { (in } \mathbb{Z})
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and with that

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The above observations can be used to embed matrices over $\mathbb{F}_{p^{s}}$ into $\mathbb{R}$, and then give guarantees on them.

## Performance comparison

## Performance Comparison (Part 1/2)



Setup: $N=6, \gamma=3 / 4$, and $Q_{\mathrm{b}}=4$.

## Performance Comparison (Part 2/2)



Setup: $N=15, \gamma=1 / 2$, and $Q_{\mathrm{b}}=4$.

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